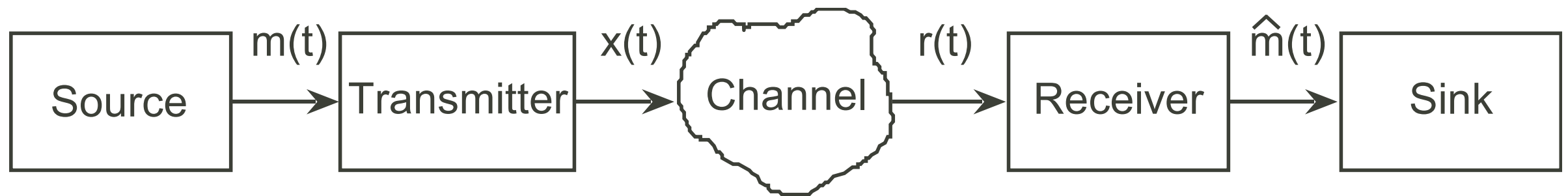


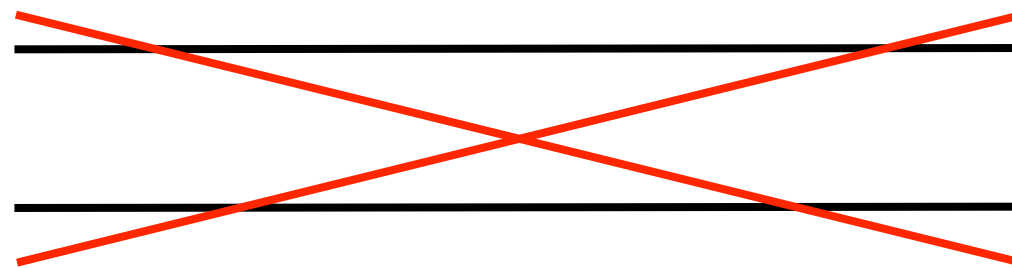
Fundamentals of Electrical Engineering

Wireline Communication Channels



Wireline Channels

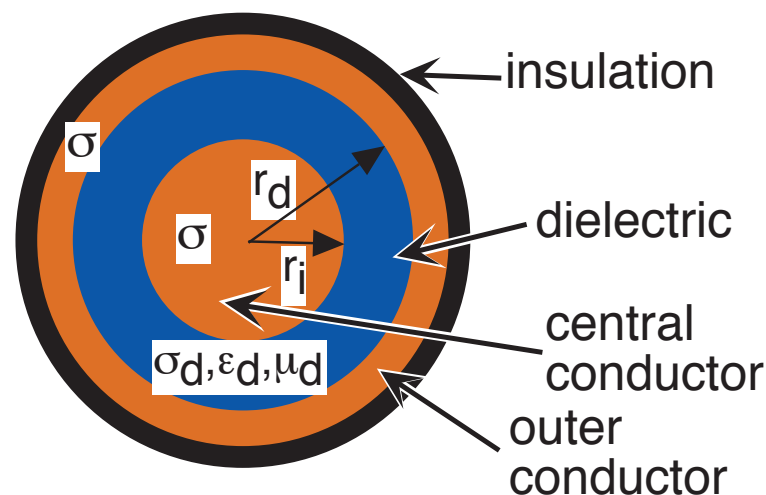
- Transmitter and receiver communicate over special wired channels
- Care must be taken to minimize *crosstalk*



two-wire pair

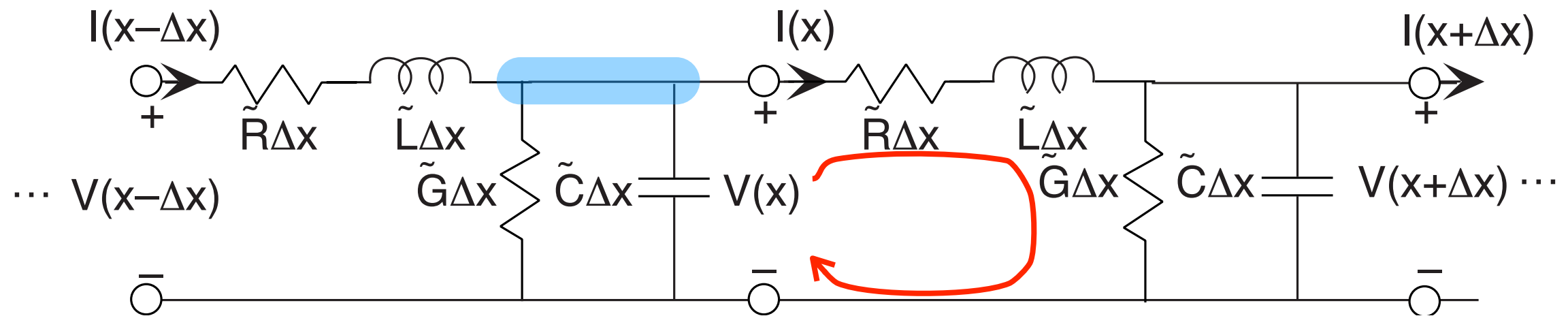


twisted pair



coaxial cable

Circuit Model for Wired Channels



KCL: $I(x) = I(x - \Delta x) - V(x) \left(\tilde{G} + j2\pi f \tilde{C} \right) \Delta x$

KVL: $V(x) = I(x) \left(\tilde{R} + j2\pi f \tilde{L} \right) \Delta x + V(x + \Delta x)$

$$\frac{d}{dx} I(x) = - \left(\tilde{G} + j2\pi f \tilde{C} \right) V(x)$$

$$\frac{d}{dx} V(x) = - \left(\tilde{R} + j2\pi f \tilde{L} \right) I(x)$$

$$\frac{d^2}{dx^2} V(x) = \left(\tilde{G} + j2\pi f \tilde{C} \right) \left(\tilde{R} + j2\pi f \tilde{L} \right) V(x)$$

Voltage Propagates!

$$\frac{d^2}{dx^2} V(x) = \left(\tilde{G} + j2\pi f \tilde{C} \right) \left(\tilde{R} + j2\pi f \tilde{L} \right) V(x)$$

$$V(x) = V_+ e^{-\gamma x} + V_- e^{+\gamma x}$$

$$\text{As } \frac{d^2}{dx^2} V(x) = \gamma^2 V(x), \quad \gamma = \pm \sqrt{\left(\tilde{G} + j2\pi f \tilde{C} \right) \left(\tilde{R} + j2\pi f \tilde{L} \right)} \\ = \pm (a(f) + jb(f))$$

$$V(x) = \begin{cases} V_+ e^{-(a+jb)x} & x > 0 \\ V_- e^{+(a+jb)x} & x < 0 \end{cases}$$

$$v(x, t) = \text{Re} \left[V_+ e^{-ax} e^{j(2\pi f t - bx)} \right], \quad x > 0$$

Voltage Propagates!

$$v(x, t) = \text{Re} \left[V_+ e^{-ax} e^{j(2\pi ft - bx)} \right], \quad x > 0$$

$$2\pi ft_2 - bx = 2\pi f(t_1 + t_2 - t_1) - bx$$

$$= 2\pi ft_1 - b \left(x - \frac{2\pi f}{b} (t_2 - t_1) \right)$$

Speed of propagation: $c = \frac{2\pi f}{b}$ Since $\lambda \cdot f = c$, $\lambda = \frac{2\pi}{b}$

$$c = \frac{2\pi f}{\text{Im} \left[\sqrt{(\tilde{G} + j2\pi f \tilde{C})(\tilde{R} + j2\pi f \tilde{L})} \right]} \xrightarrow{f \rightarrow \infty} \frac{1}{\sqrt{\tilde{L}\tilde{C}}}$$

$$\text{and } a = \text{Re} \left[\sqrt{(\tilde{G} + j2\pi f \tilde{C})(\tilde{R} + j2\pi f \tilde{L})} \right] \xrightarrow{f \rightarrow \infty} \frac{1}{2} \left(\frac{\tilde{R}}{Z_0} + \tilde{G} Z_0 \right)$$

Current Propagates, too!

$$V(x) = V_+ e^{-\gamma x} + V_- e^{+\gamma x}$$

$$\text{As } \frac{d}{dx} V(x) = - \left(\tilde{R} + j2\pi f \tilde{L} \right) I(x)$$

$$\text{and } \gamma = \sqrt{\left(\tilde{G} + j2\pi f \tilde{C} \right) \left(\tilde{R} + j2\pi f \tilde{L} \right)}$$

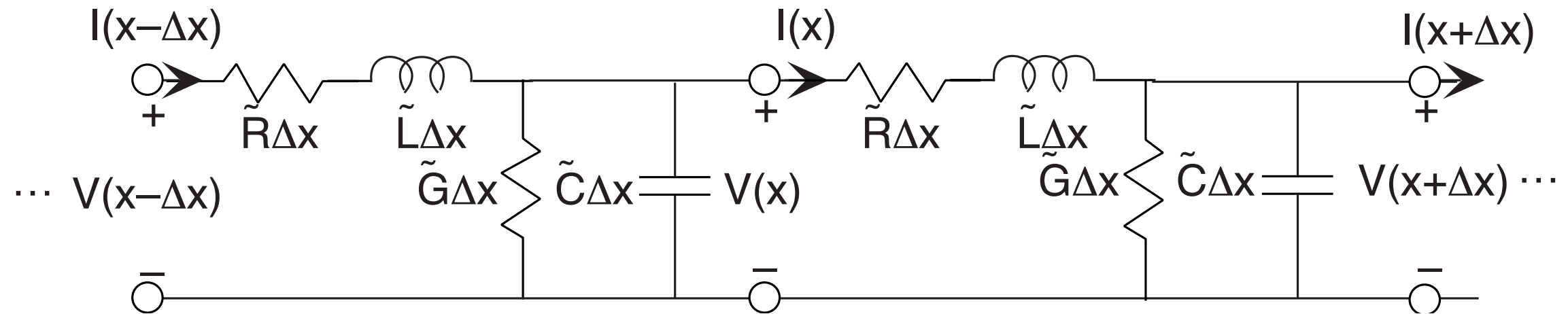
$$I(x) = \sqrt{\frac{\tilde{G} + j2\pi f \tilde{C}}{\tilde{R} + j2\pi f \tilde{L}}} V_+ e^{-\gamma x} - \sqrt{\frac{\tilde{G} + j2\pi f \tilde{C}}{\tilde{R} + j2\pi f \tilde{L}}} V_- e^{+\gamma x}$$

$$v(x, t) = \text{Re} \left[V_+ e^{-ax} e^{j(2\pi ft - bx)} \right], \quad x > 0$$

$$i(x, t) = \text{Re} \left[I_+ e^{-ax} e^{j(2\pi ft - bx)} \right], \quad x > 0$$

$$I_+ = \sqrt{\frac{\tilde{G} + j2\pi f \tilde{C}}{\tilde{R} + j2\pi f \tilde{L}}} V_+$$

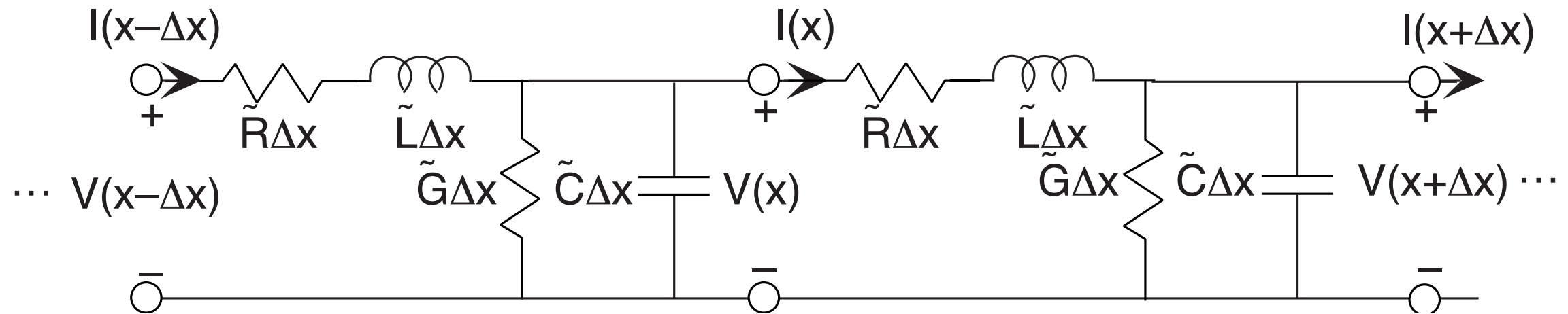
Wireline Channels



$$v(x, t) = \text{Re} \left[V_+ e^{-ax} e^{j(2\pi ft - bx)} \right], \quad x > 0$$

At sufficiently high frequencies, signals propagate with very little loss at speed $\frac{1}{\sqrt{\tilde{L}\tilde{C}}}$

Wireline Channels

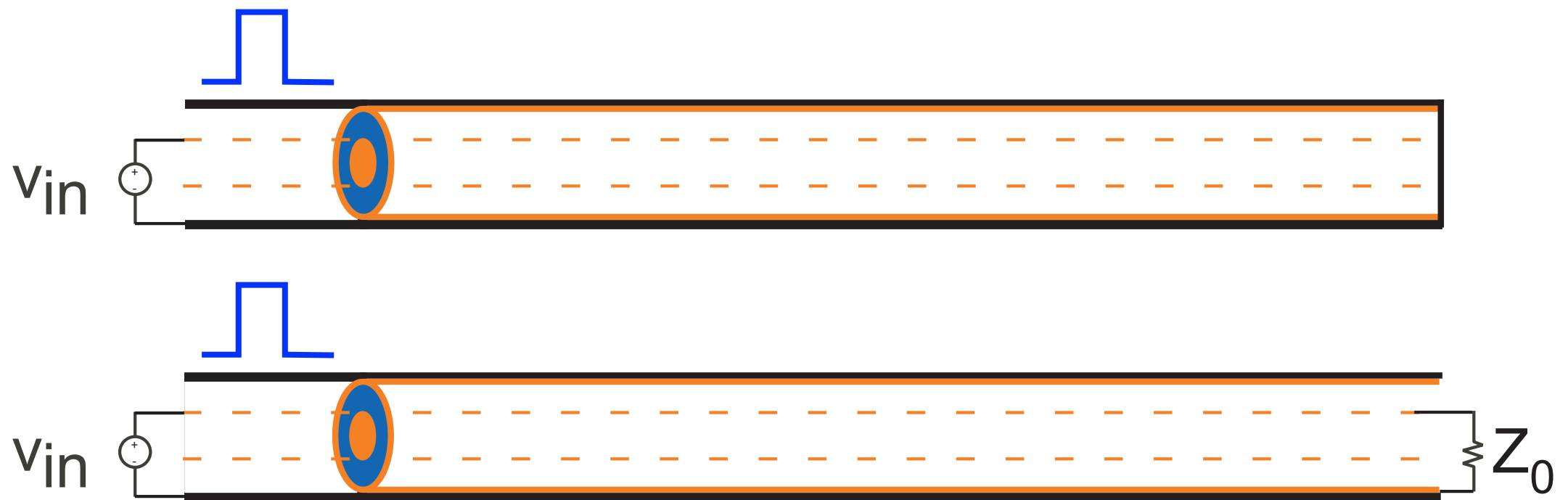


$$\frac{V(x)}{I(x)} = \sqrt{\frac{\tilde{R} + j2\pi f \tilde{L}}{\tilde{G} + j2\pi f \tilde{C}}} \equiv Z_0 \text{ (characteristic impedance)}$$

$$\lim_{f \rightarrow \infty} Z_0 = \sqrt{\frac{\tilde{L}}{\tilde{C}}}$$

Wireline Channels

- An important detail is that the circuit model assumed an *infinitely* long cable
- Unless you *terminate* a cable properly, reflections will occur



$$Z_0 = 50 - 75 \, \Omega$$

Wireline Channels

- Designed to minimize interference
- Receiver and transmitter must be connected to the cable
- Well-designed cable has little attenuation at high frequencies
- At low frequencies, amplitude decreases *exponentially* with distance

Wireless vs. Wireline

- Flexible connections via wireless, but communication distance limited by power and “natural” considerations
- Wireline requires a direct connection between transmitter and receiver
- Wireless prone to much more noise and interference